

# High-Power Considerations in Metamaterial Antennas

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14. ABSTRACT  Many metamaterial types rely on resonant behaviors that produce high fields within their structures. However, if a metamaterial can operate away from resonance (e.g., low-index or zero-index metamaterials), it can be well-suited for HPM applications. Artificial Magnetic Conducting surfaces often exhibit high field enhancement at resonance with unoptimized MFEFs over 30. Genetic algorithm optimization was successfully employed to design single- and dual-band AMC surfaces with 50% reduced MFEF for HPM applications.					
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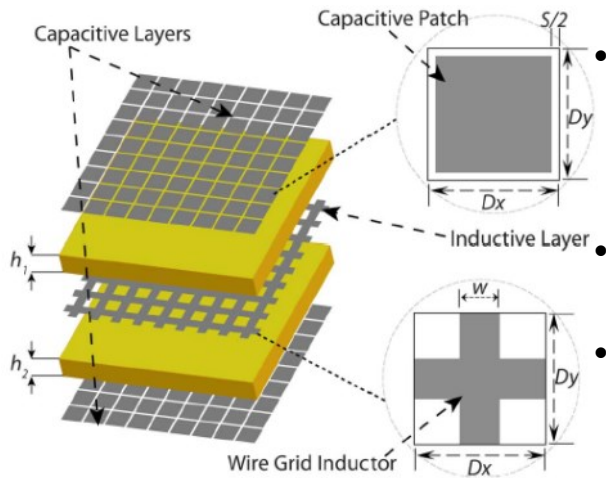


# Overview

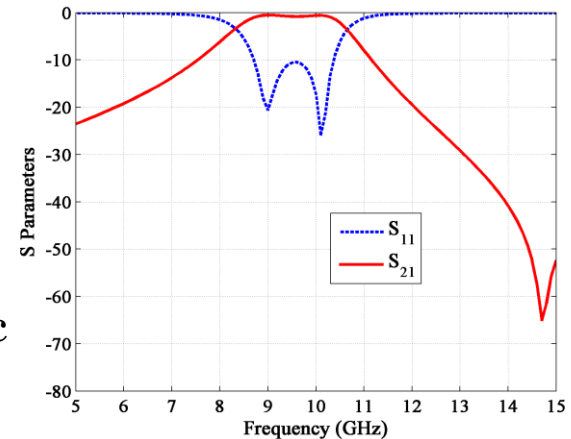
1. Frequency Selective Surfaces and Metamaterials under High-Power Microwaves
2. Artificial Magnetic Conducting Surfaces
3. Periodic FE-BI Analysis
4. AMC Synthesis using Genetic Algorithms
5. Design of Single-Band and Dual-Band AMC Surfaces
6. Summary & Conclusion

# Frequency Selective Surfaces for High-Power Microwave (HPM) Applications

## Second-order Bandpass Filter with Low-profile Frequency Selective Surfaces (FSS)



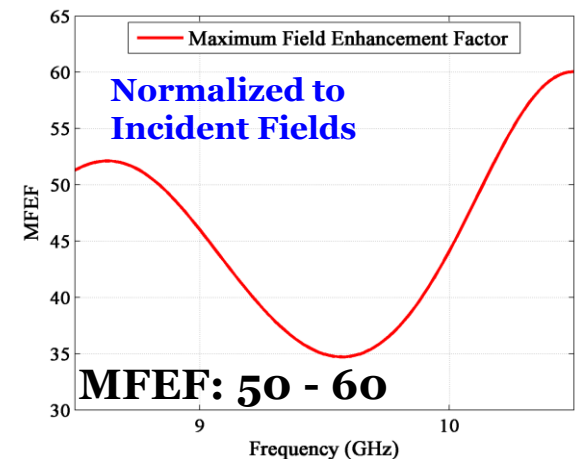
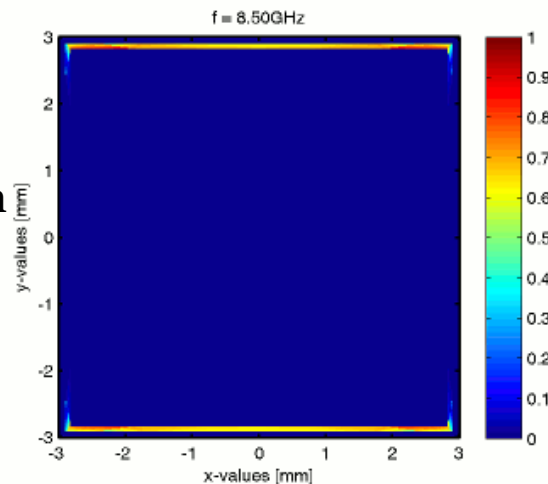
- Multi-layer FSS with non-resonant constitutive elements (metallic patches and wire grids).
- Verified filtering properties by HFSS simulations.
- Identified maximum electric field in the gap regions between capacitive patches.



M. Al-Joumayly, and N. Behdad, "A New Technique for Design of Low-Profile, Second-Order, Bandpass Frequency Selective Surfaces," *IEEE Transactions on Antennas and Propagation*, Vol. 57, pp. 452-459, 2009.

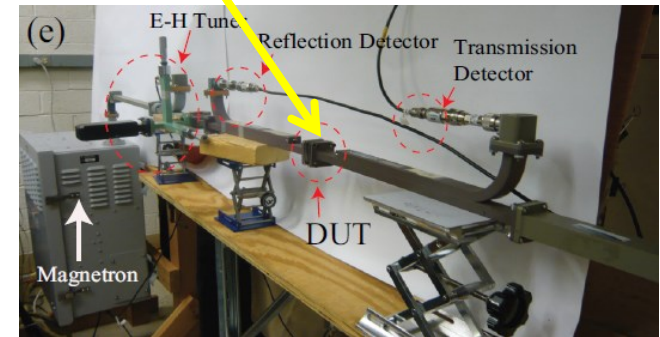
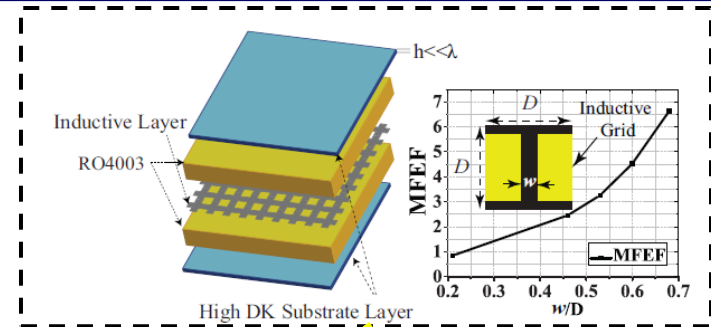
**Evaluated the electric field distributions on the top capacitive patch surface at different frequencies.**

- Implemented numerical models based on full-wave simulations to effectively evaluate the electric field distributions within the metamaterial and FSS.
- The simulation results can be used to determine the power handling capability of such structures.

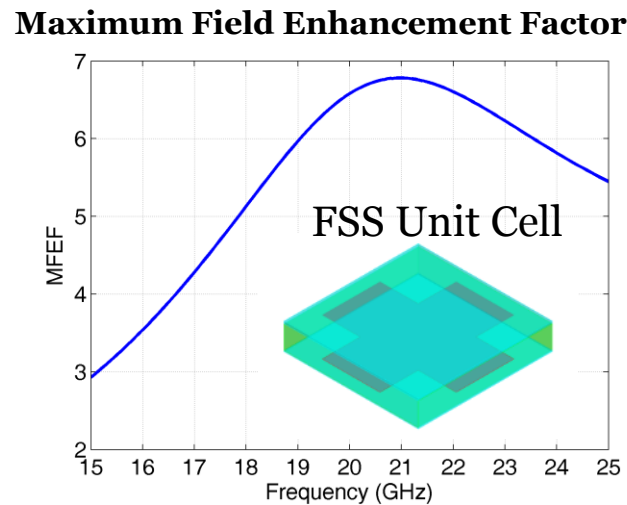
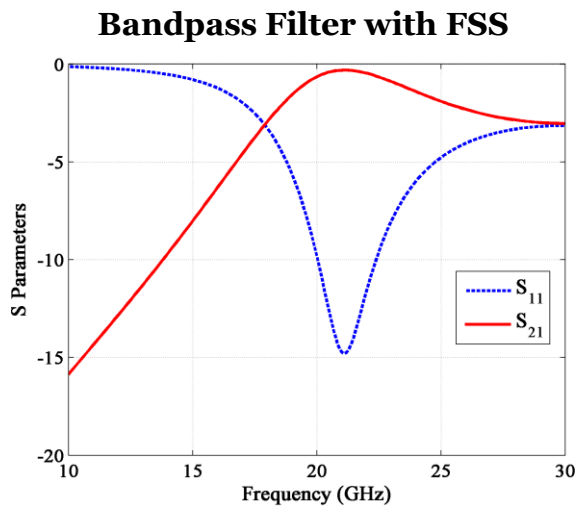


# Frequency Selective Surfaces for High-Power Microwave (HPM) Applications

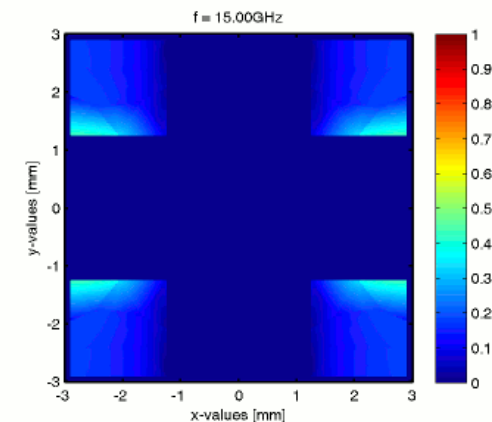
- Replacing the capacitive patches with high permittivity (*e.g.*  $\epsilon_r=20$ ) dielectric layers can reduce the maximum electric field enhancement factor.
- Peak power of 25 kW (power density of  $1.08 \times 10^8 \text{ W/m}^2$ ) was experimentally demonstrated.
- Both the filtering performance and field enhancement factor were verified by our full-wave numerical simulations.



L. Meng, and N. Behdad, "Frequency Selective Surfaces for High-Power Microwave (HPM) Applications," *IEEE AP-S*, 2012.

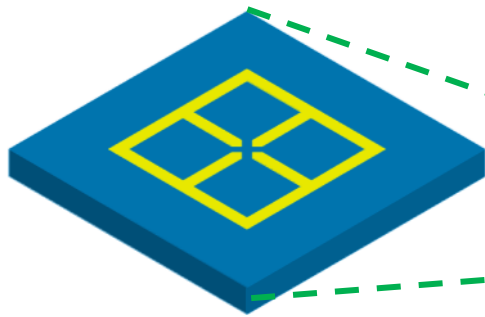


## Electric Field Distributions on the Inductive Patch Surface

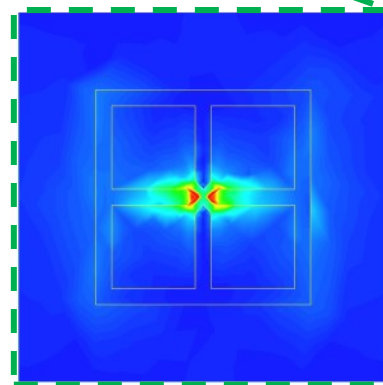


# Metamaterials for High-power RF Applications

- Metamaterials with unit cells that support a confined resonance can have significantly stronger electric fields than the incident wave, which can cause voltage breakdown in air or dielectrics.
- HPM electromagnetic fields can induce losses and heating on the unit cells of metamaterials, which cause melting, structural deformation and variation in the responses.

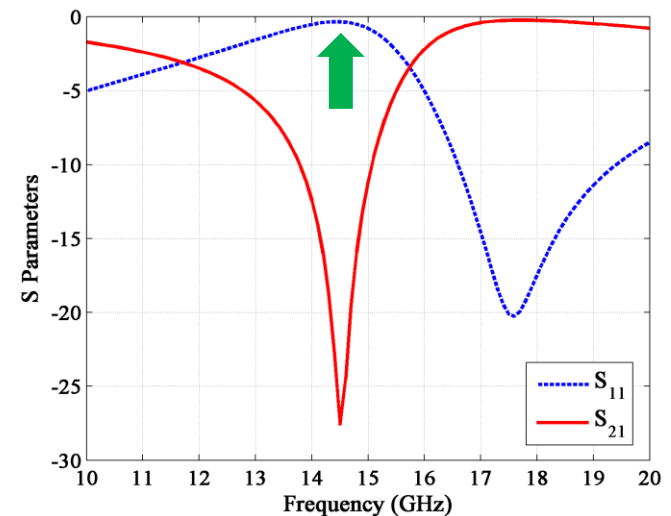


An electric LC resonator metamaterial unit cell composed of metallic traces on top of a dielectric substrate.



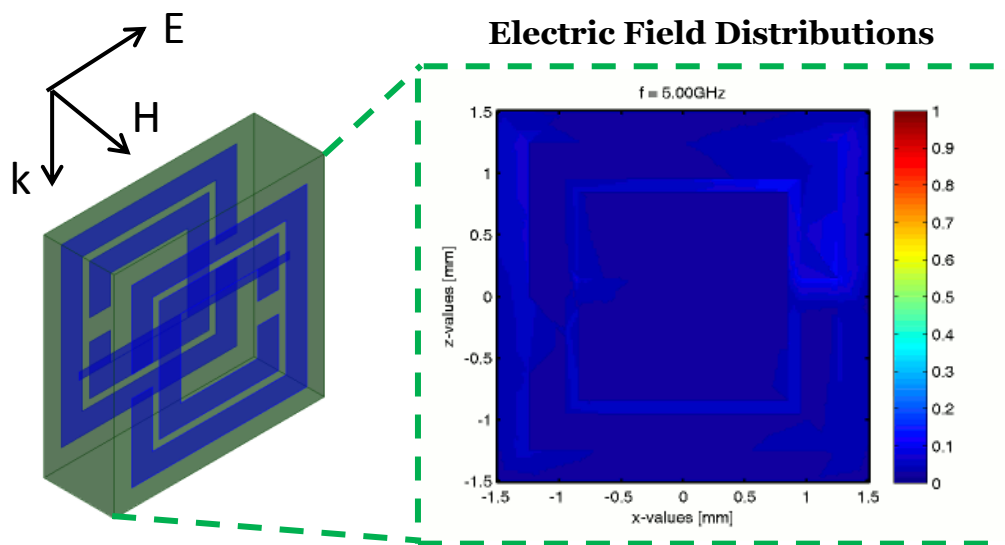
**Maximum Field  
Enhancement Factor  
of 90 at 14.6 GHz**

The metamaterial supports a strong LC resonance due to the gap capacitance and loop inductance, which gives rise to highly enhanced electric fields in the middle of the unit cell.

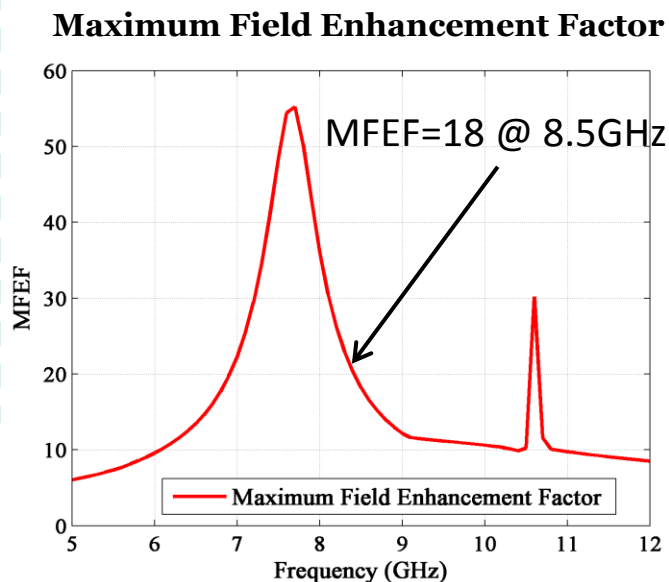
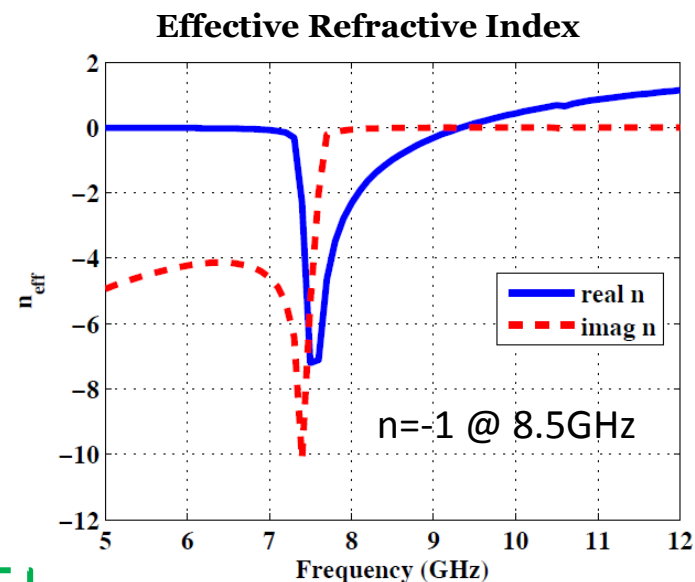


# High-power Considerations for Negative Index Metamaterials

- A representative metamaterial, the negative index material (NIM), has been used to implement flat lenses with advanced focusing properties.
- Typical NIM designs exhibit high absorption losses and high field enhancement near the resonance, limiting their application for HPM.



A negative index material composed of split ring resonators and metallic strips. The maximum field occurs at the surface of the split ring resonators.

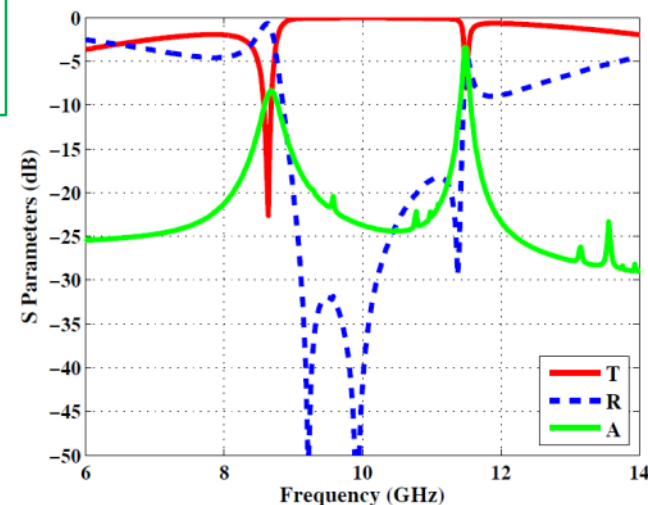
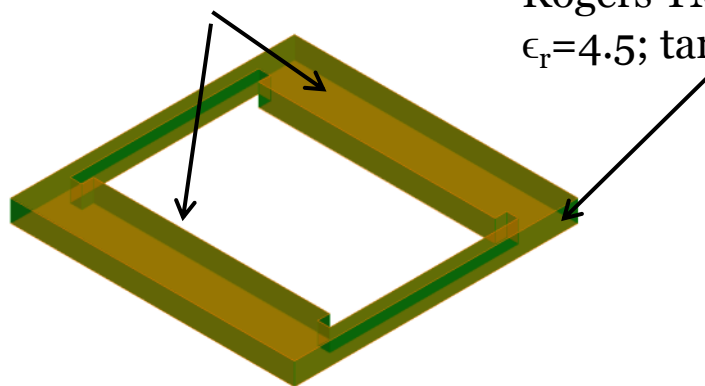


# Low Loss Negative Index Materials for High-power RF Applications

A negative index metamaterial design with low loss was realized by using powerful global optimization techniques.

Copper Layers  
Thickness = 0.035mm

Rogers TMM4  
 $\epsilon_r=4.5$ ;  $\tan\delta=0.002$

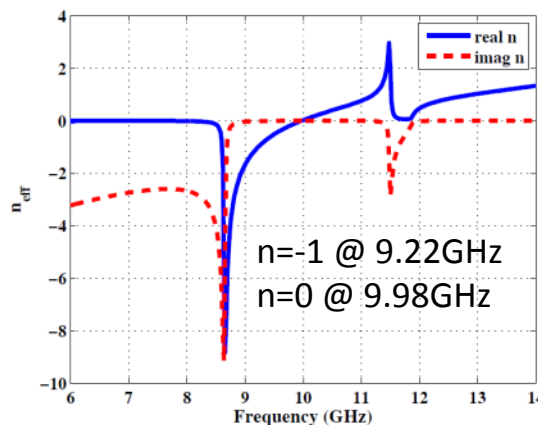


**Achieved reduced absorption and reflection loss by optimizing the NIM structures.**

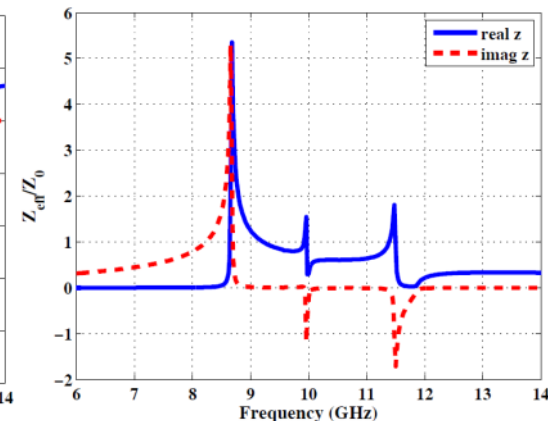
Modified Fishnet with Notches



Effective Refractive Index



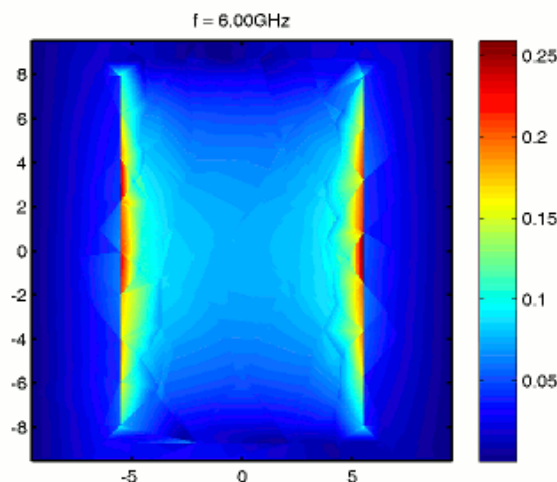
Effective Impedance



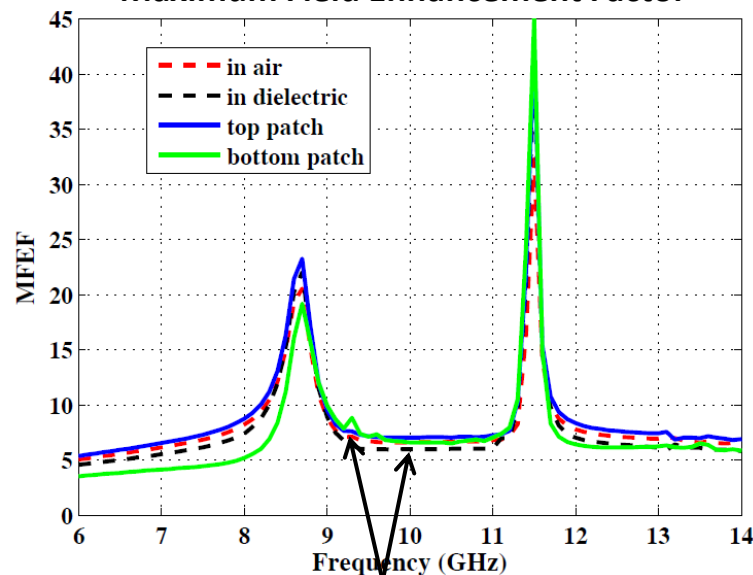


# Low Loss Negative Index Materials for High-power RF Applications

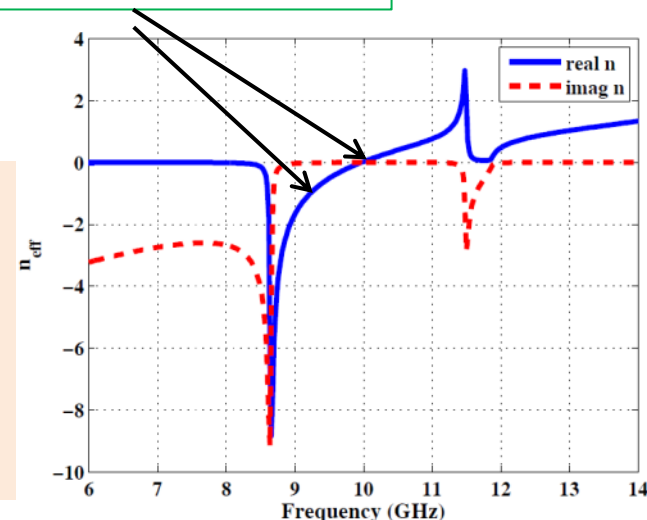
Maximum  $E$  field occurs at the top metallic patch layer (blue curve)



Maximum Field Enhancement Factor



MFEF=7.6;  $n=-1$  @ 9.22GHz  
MFEF=7.0;  $n=0$  @ 9.98GHz

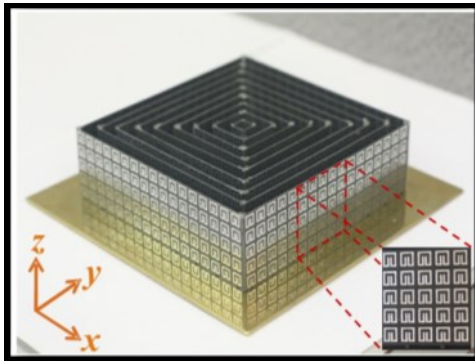


- Reduced overall field enhancement with optimized NIM design in the negative and zero index bands.
- At NIM band ( $n=-1$ ), the MFEF reduced from 18 to 7.6 comparing to the previous NIM design with SRRs.
- At ZIM band ( $n=0$ ), MFEF reduces from 11.5 to 7.0.

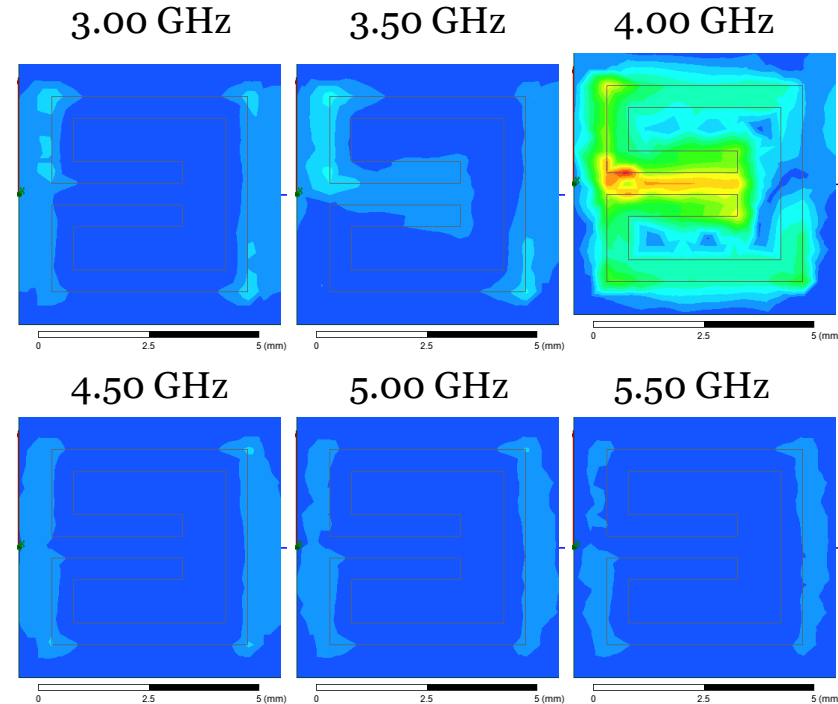
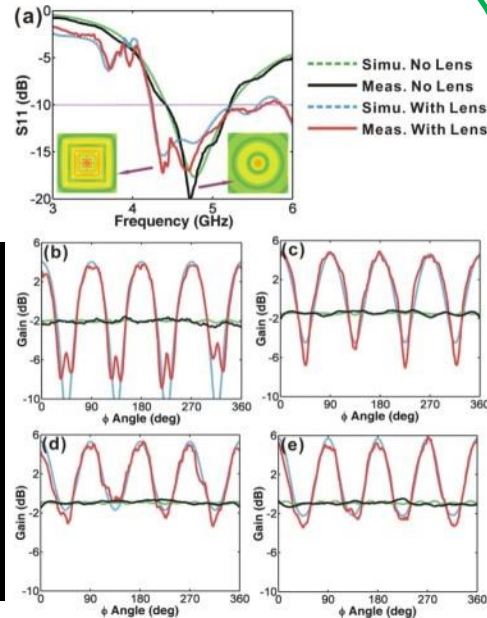
# Zero- or Low-Index Metamaterial for High-power RF Applications

## Quadbeam Lens with ZIM/LIM Metamaterial

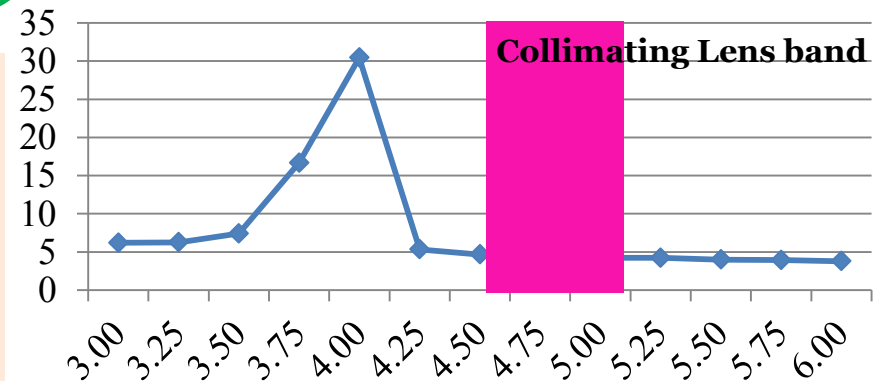
The impedance and pattern bandwidths of the feed dipole were increased by adding the lens.



Z. H. Jiang, M. D. Gregory, and D. H. Werner, "Experimental Demonstration of a Broadband Transformation Optics Lens for Highly Directive Multibeam Emission," *Phys. Rev. B*, 84, 165111 (2011).



## Maximum Field Enhancement Factor



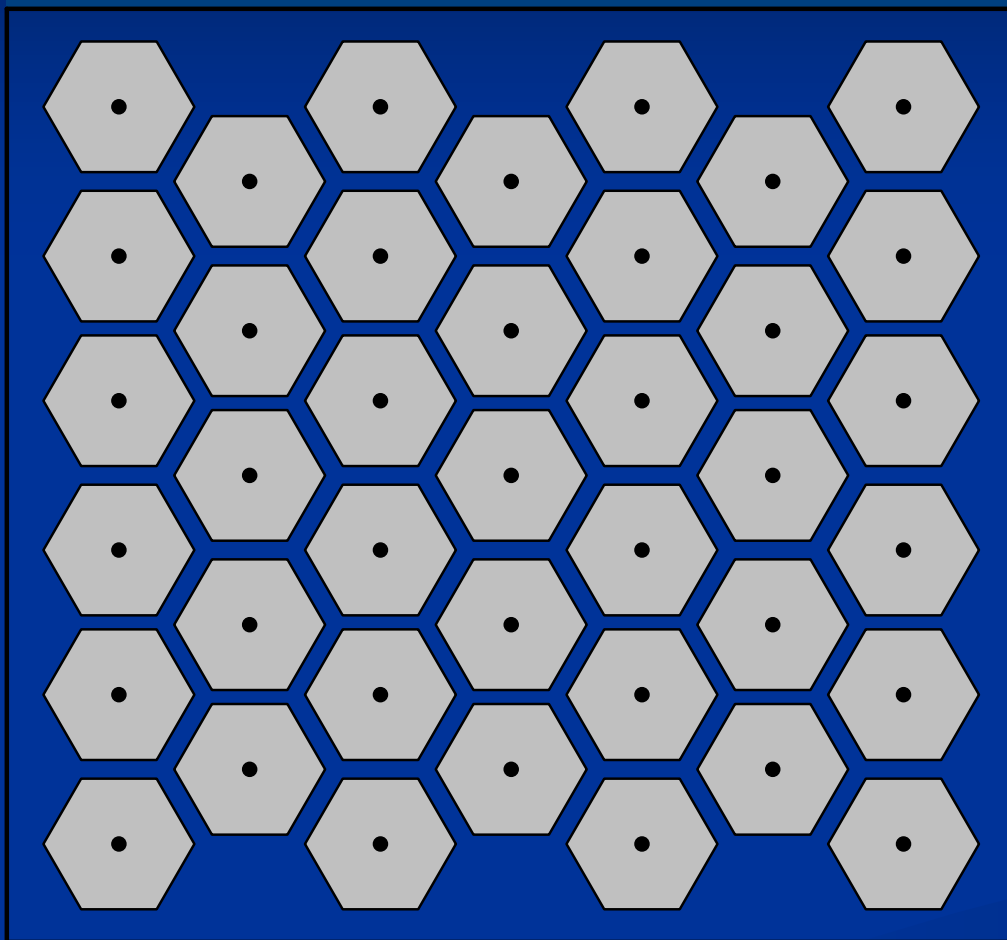
- Near-zero index metamaterials can be used to implement lenses for directive radiation.
- Full-wave simulations reveal that the field intensity can be reduced by operating away from resonance, such as in the near-zero index band, suitable for HPM applications.

# Artificial Magnetic Conductors

Cross section of a Sievenpiper's high-impedance surface



Top view of the Sievenpiper's high-impedance surface



## Properties:

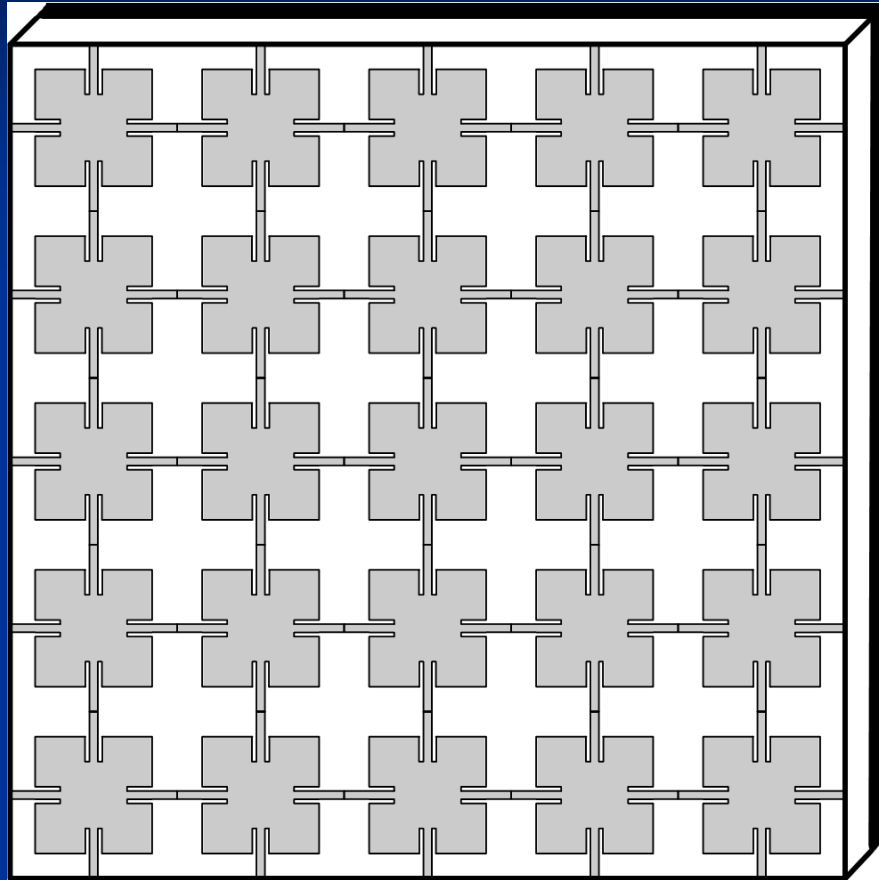
- All metallic, two layered, connected with vias.
- High-impedance surface suppresses surface waves at the forbidden frequency range

## Sheet impedance

$$Z = \frac{j\omega L}{1 - \omega^2 LC} \quad \omega_0 = \frac{1}{\sqrt{LC}}$$

D. Sievenpiper, L. Zhang, Romulo F. Jimenez Broas, N. G. Alexopolous and Eli Yablonovitch, "High-Impedance Electromagnetic Surfaces with a Forbidden Frequency Band," *IEEE Transactions on Microwave Theory and Techniques*, Vol. 47, No. 11, pp. 2059 – 2074, Nov. 1999.

# Artificial Magnetic Conductors



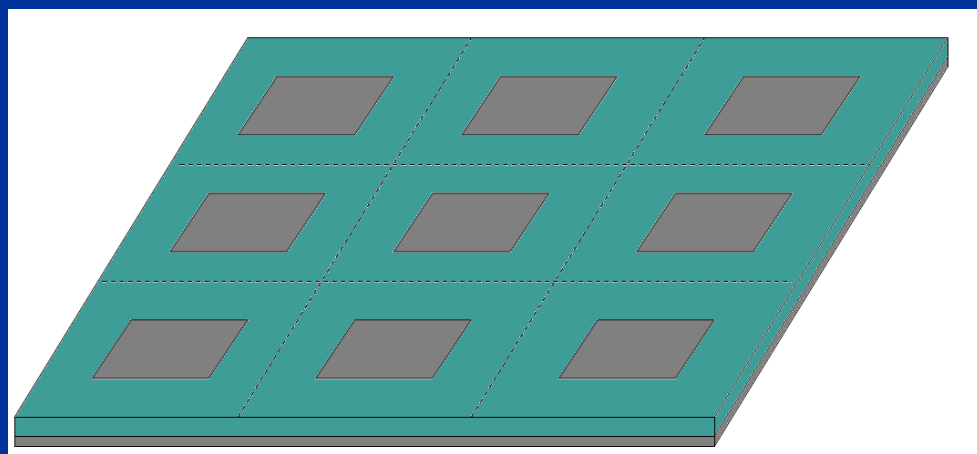
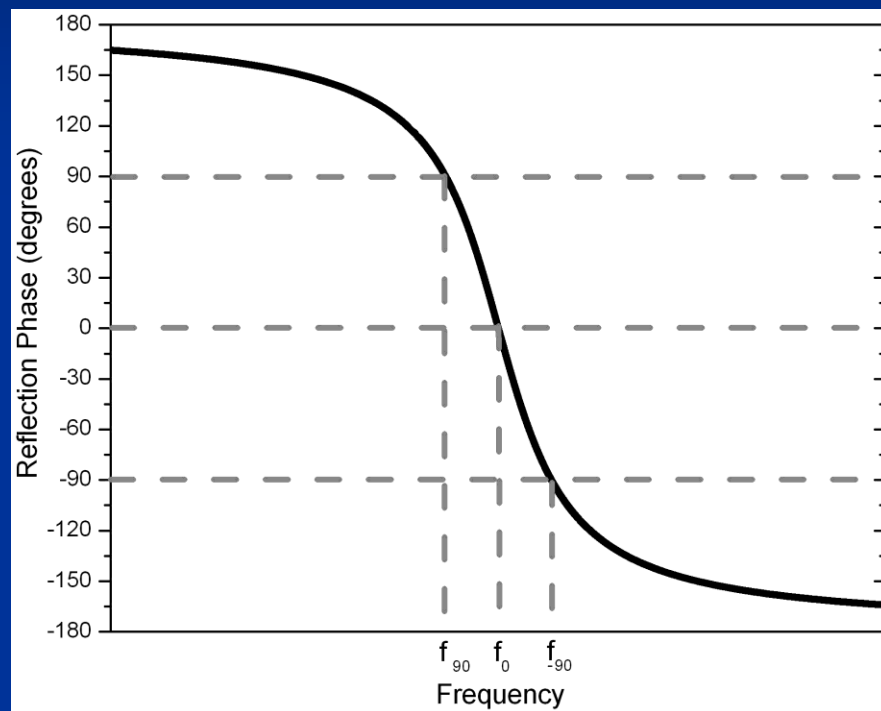
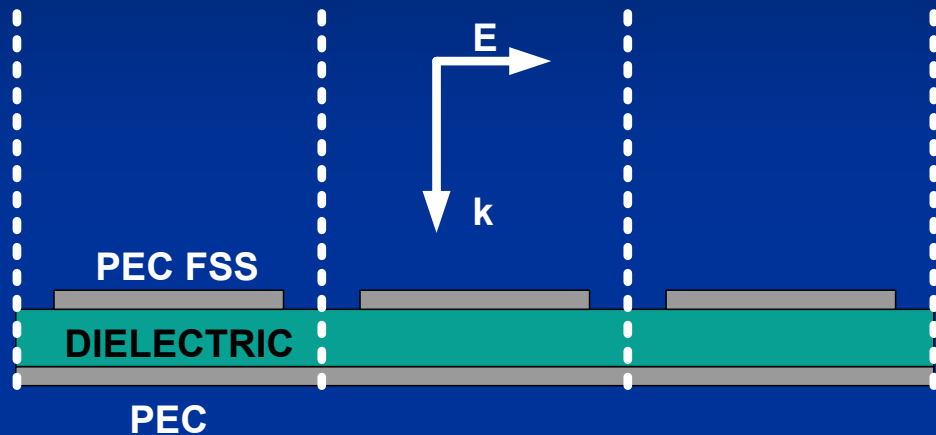
Top view of a Itoh's planar EBG surface

## Properties:

- Metallo-dielectric.
- Doubly periodic Frequency Selective Surface on top of a dielectric backed by a conducting ground plane.
- If optimized; multiband (fractal), and magnetically loaded designs possible

K-P. Ma, K. Hirose, F-R. Yang, Y. Qian, and T. Itoh, "Realisation of magnetic conducting surface using novel photonic bandgap structure," *IEE Electronics Letters*, Vol. 34, No. 21, pp. 2041-2042, 15th October 1998.

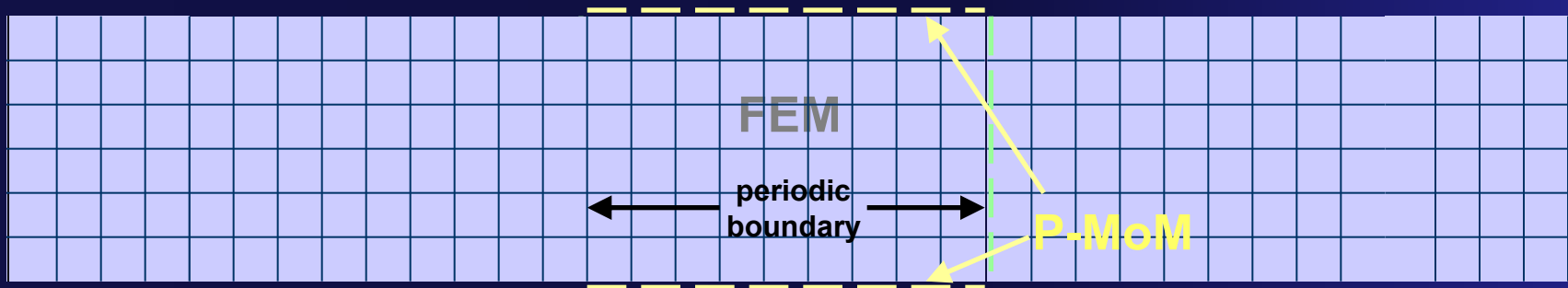
# Artificial Magnetic Conductors



Three unitcells by three unitcells view



# Periodic FE-BI Analysis



- Periodic BC  $\rightarrow$  Simulate one unit cell
- Interior region: Periodic Finite Element Method (P-FEM)

$$F(E_{\text{ad}}, E) = \iiint_V \left[ \frac{1}{\mu_r} (\nabla \times E_{\text{ad}}) \cdot (\nabla \times E) - k_0^2 \epsilon_r E_{\text{ad}} \cdot E \right] dv + jk_0 Z_0 \iint_S E_{\text{ad}} \cdot (H \times n) ds$$

- Top & bottom boundary: Periodic Method of Moments (P-MoM)

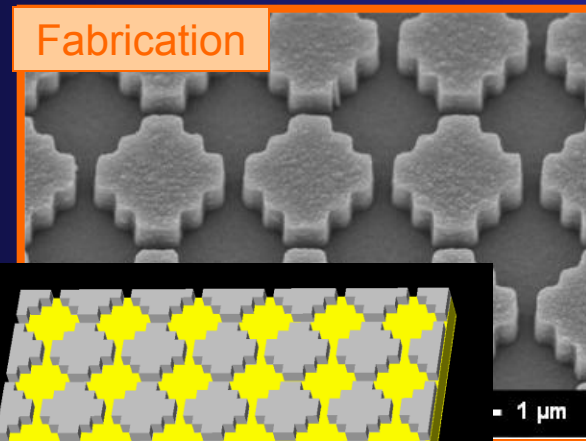
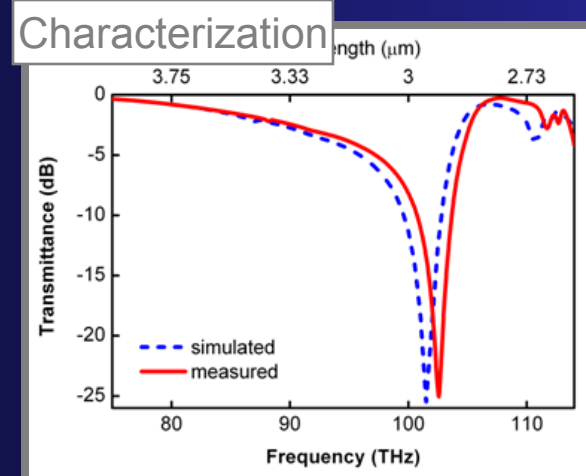
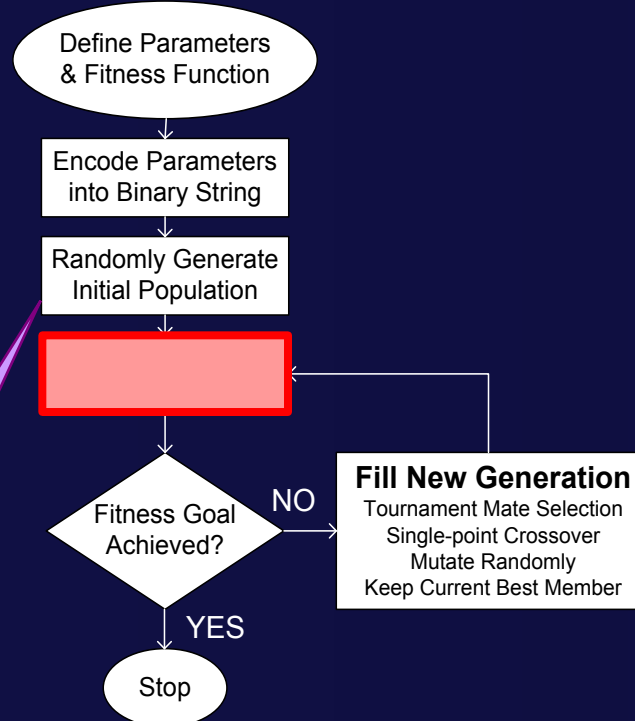
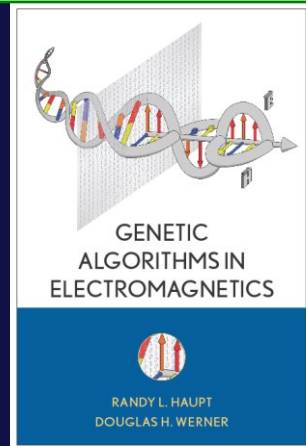
$$H = -2j \frac{1}{Z_0} \left[ \iint_S G_p(r, r_s) (E \times n) ds + \frac{1}{k_0^2} \nabla \iint_S G_p(r, r_s) \nabla_s \cdot (E \times n) ds \right] + H^{\text{exc}}$$

- Accelerated computation for  $G_p(r, r_s)$  using Ewald transformation

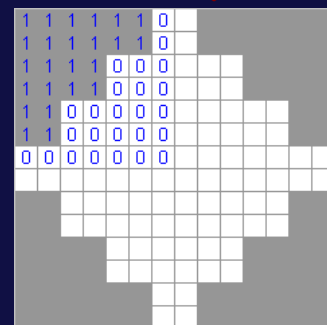
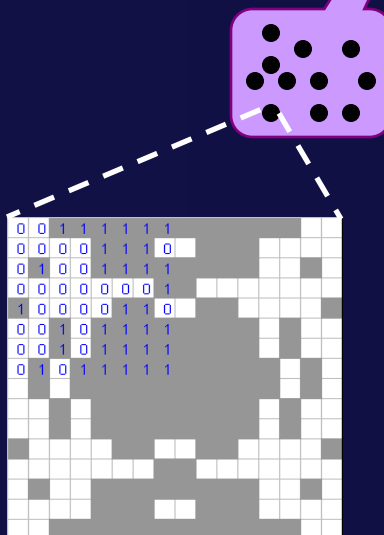
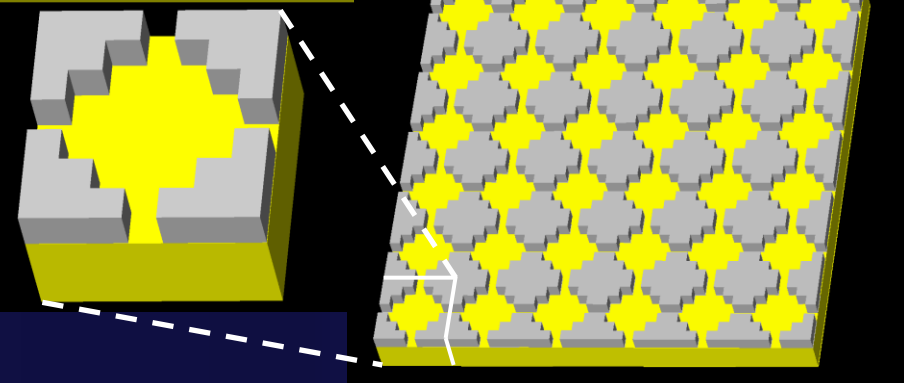


# GA Design Approach

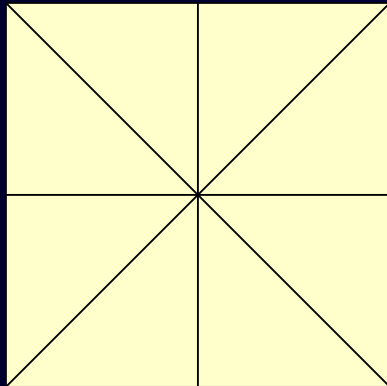
Haupt and Werner (2007)



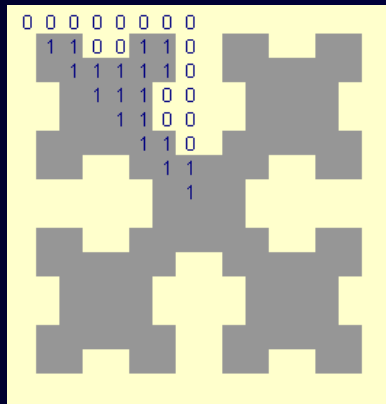
**Optimized Design**



# GA EBG Synthesis



**8-Fold Symmetry**



**Cell Size**

## • Implementation of GA

- **8-fold symmetry** applied to metal screen pattern to achieve polarization insensitivity
- Cell size and other EBG parameters are encoded in the chromosome
- Each sample is evaluated for fitness against the ideal frequency response

$$Cost = \sum_{freqs} \{\varphi_R - 0.0\}^2$$

$$Cost = \sum_{freqs} \{\varphi_R - 0.0\}^2 + \{\max(|E|) - 0.0\}^2$$

**Fitness Function**

FSS Cell Geometry Applying 8-Fold Symmetry | **Cell Size** | Other Parameters  
 00000000|1100110|111110|11100|1100|110|11|1 | **01101110** | ...



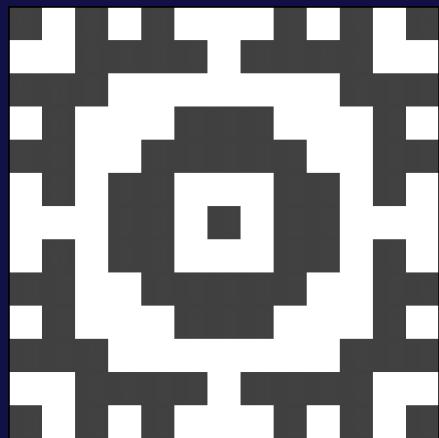


# Single-Band AMC Optimization

Design Goal: AMC Condition at 1.25 GHz

$$Cost = \sum_{freqs} \{\varphi_R - 0.0\}^2$$

$$freqs = \{1.25 \text{ GHz}\}$$



5.0 cm

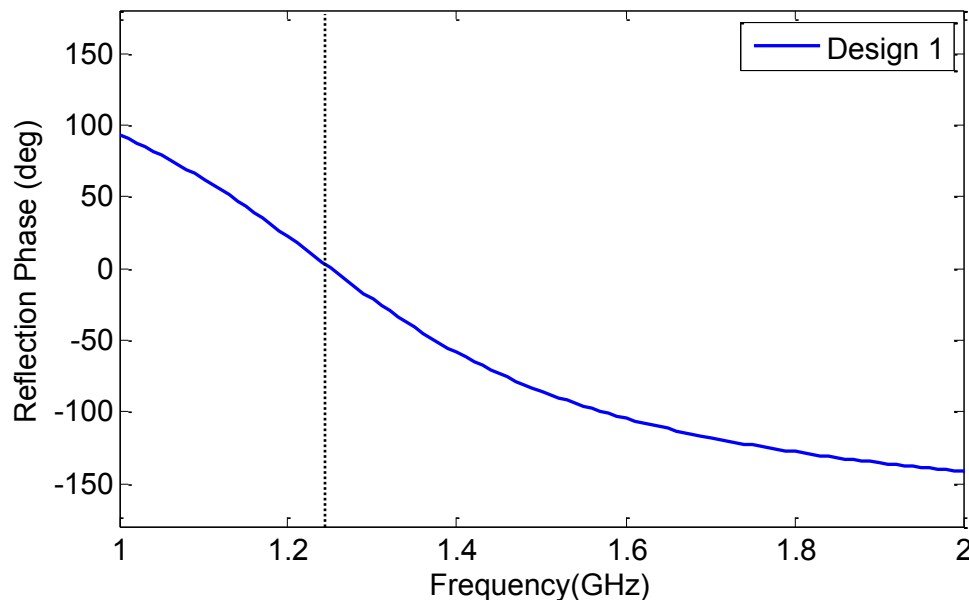
FSS Screen



PEC Ground



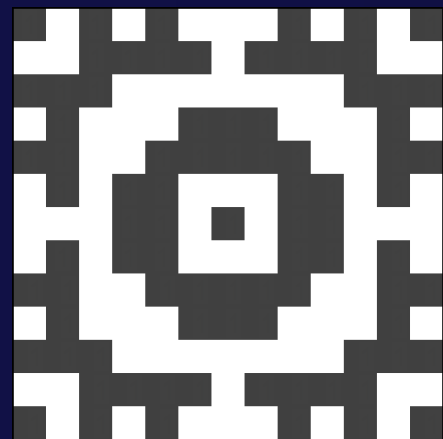
1.75 cm





# Single-Band AMC Field Enhancement

Peak MFEF of 34.7 @ 1.25 GHz.



5.0 cm

FSS Screen



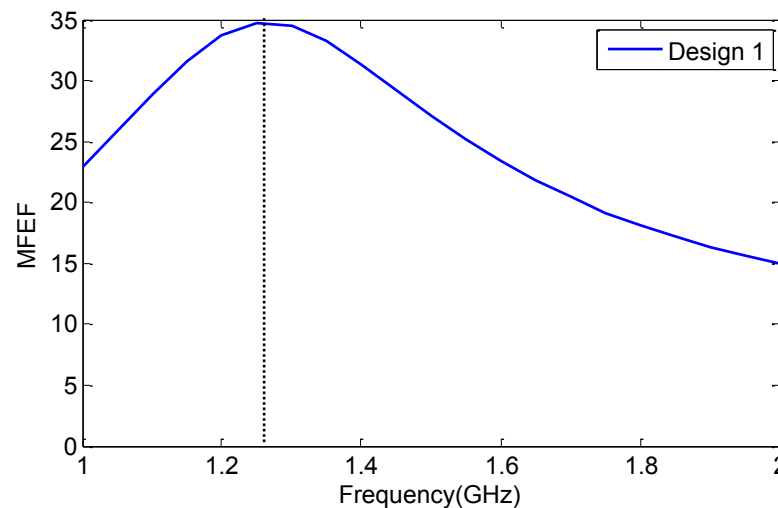
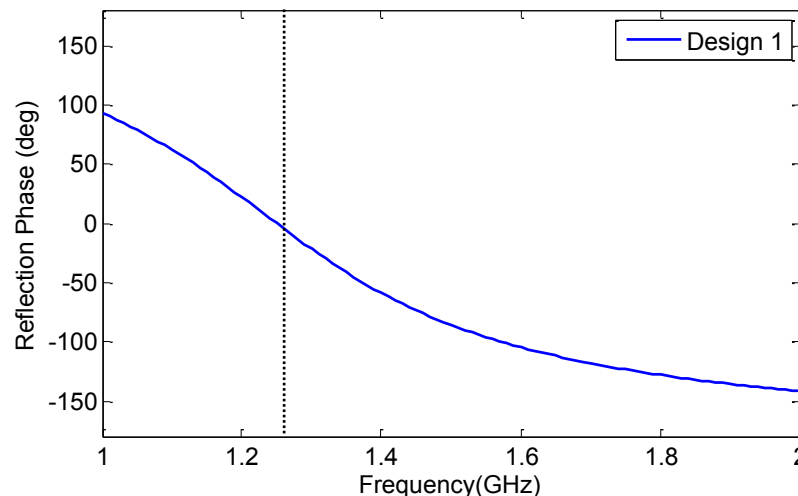
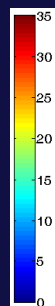
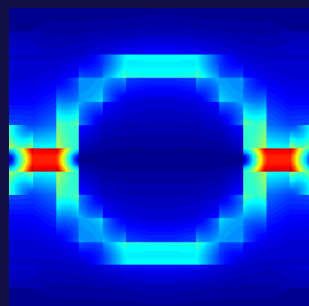
$\epsilon_r = 2.0$

PEC Ground



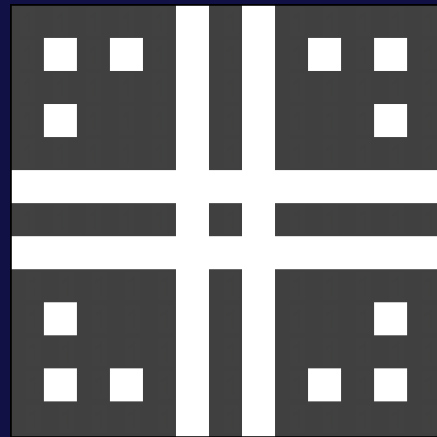
1.75 cm

$|E|$  Enhancement @ 1.25 GHz





# Single-Band AMC With Suppressed MFEF



4.86 cm

FSS Screen

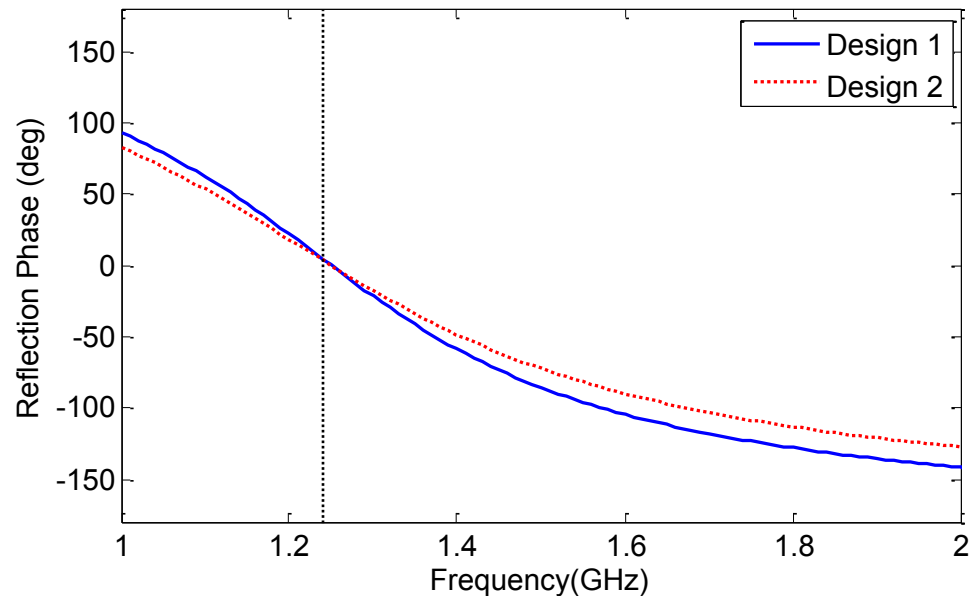
$\epsilon_r = 2.0$  2.0 cm

PEC Ground

Design Goal: AMC Condition at 1.25 GHz

$$Cost = \sum_{freqs} \{\varphi_R - 0.0\}^2 + \{\max(|E|) - 0.0\}^2$$

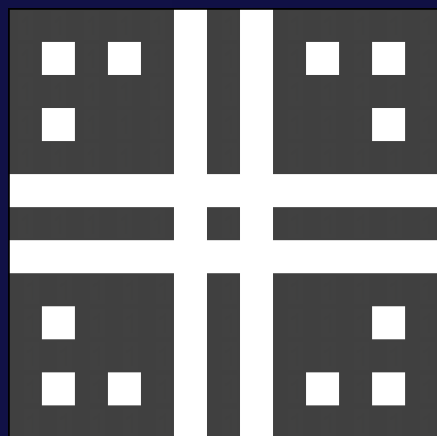
$freqs = \{1.25 \text{ GHz}\}$





# Single-Band AMC MFEF Comparison

Peak MFEF @ 1.25 GHz Reduced from 34.7 to 14.3



4.86 cm

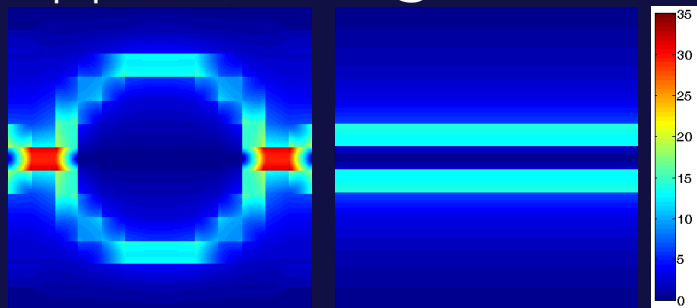
FSS Screen

$\epsilon_r = 2.0$

PEC Ground

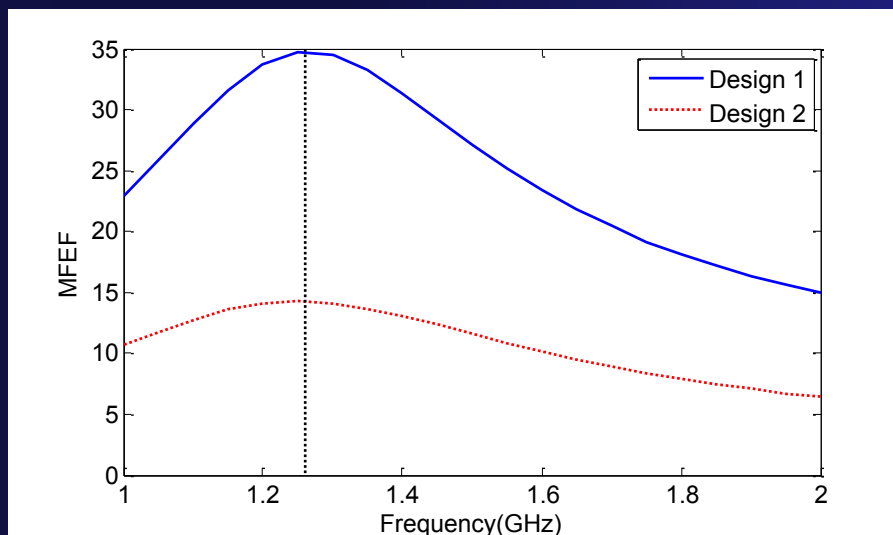
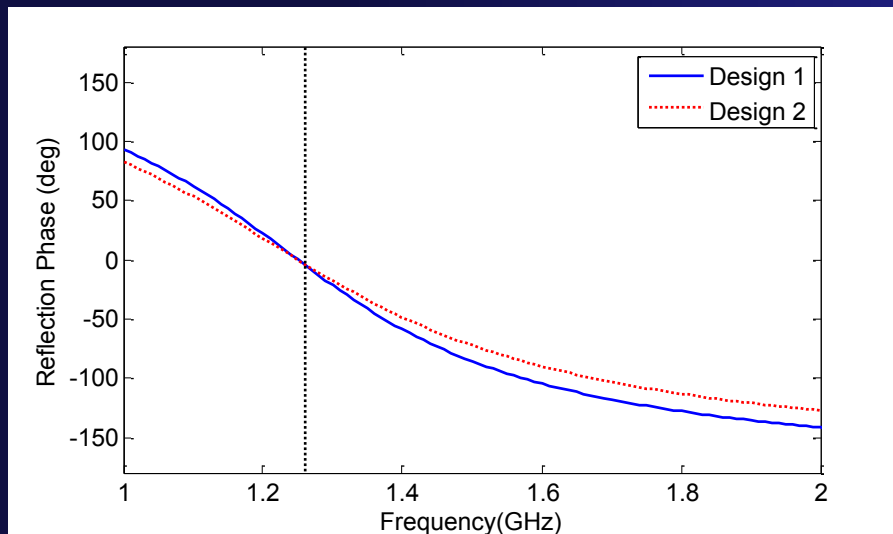
2.0 cm

$|E|$  Enhancement @ 1.25 GHz



Design 1

Design 2



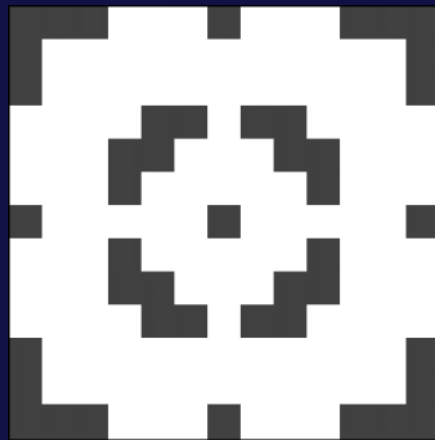


# Dual-Band AMC Optimization

Design Goal: AMC Condition at 1.25 GHz and 1.75 GHz

$$Cost = \sum_{freqs} \{\varphi_R - 0.0\}^2$$

$$freqs = \{1.25 \text{ GHz}, 1.75 \text{ GHz}\}$$

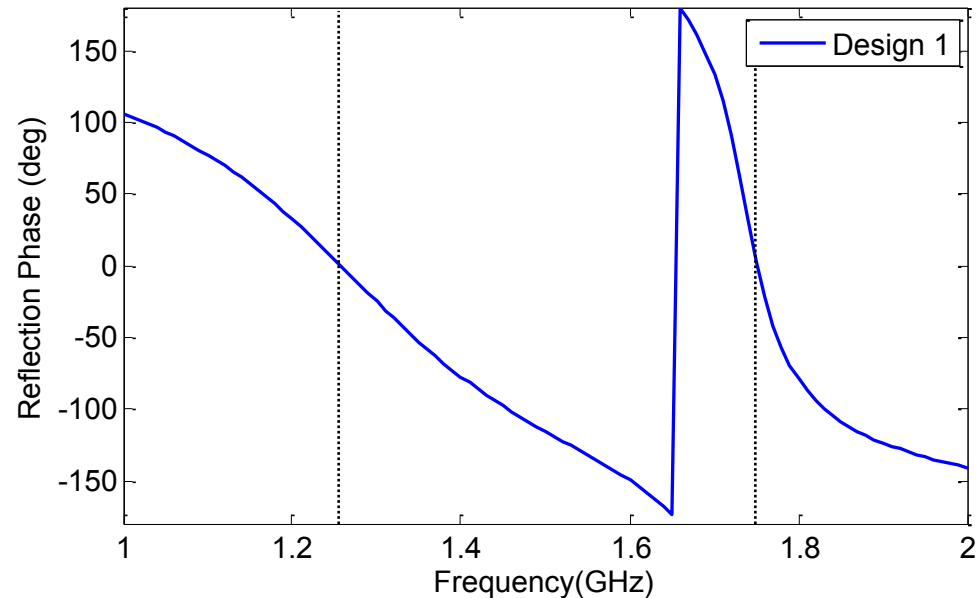


7.09 cm

FSS Screen

$\epsilon_r = 7.76$  1.52 cm

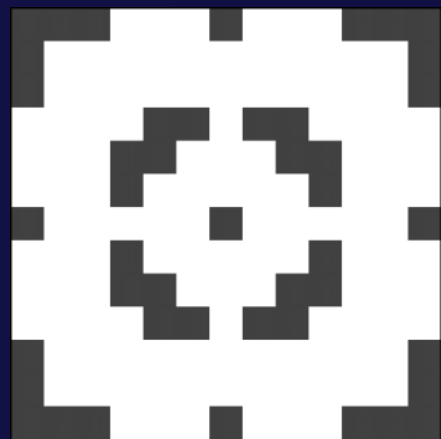
PEC Ground





# Dual-Band AMC Field Enhancement

Peak MFEF of 31.0 @ 1.75 GHz.



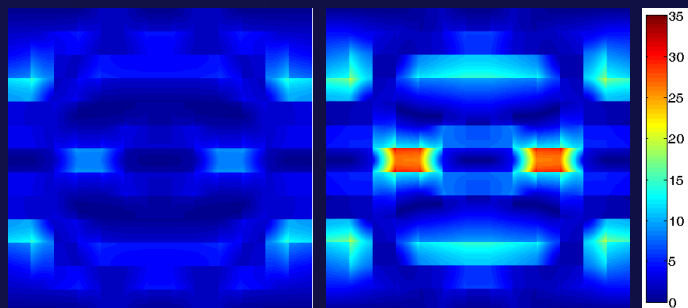
7.09 cm

FSS Screen

$\epsilon_r = 7.76$  1.52 cm

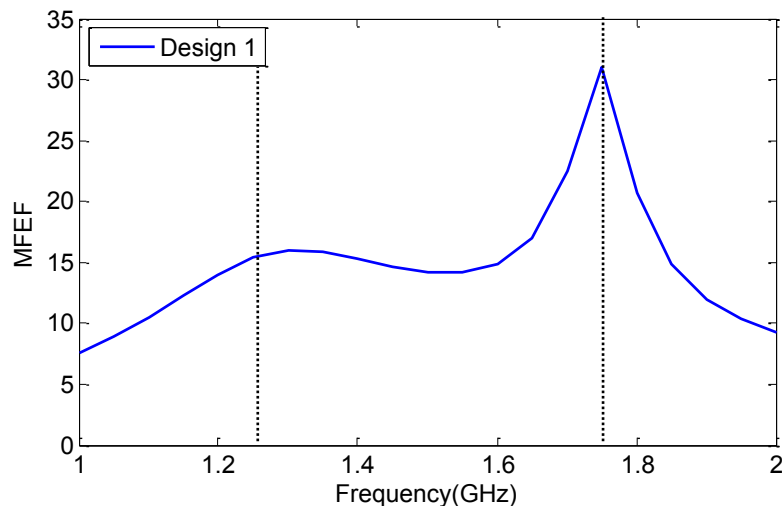
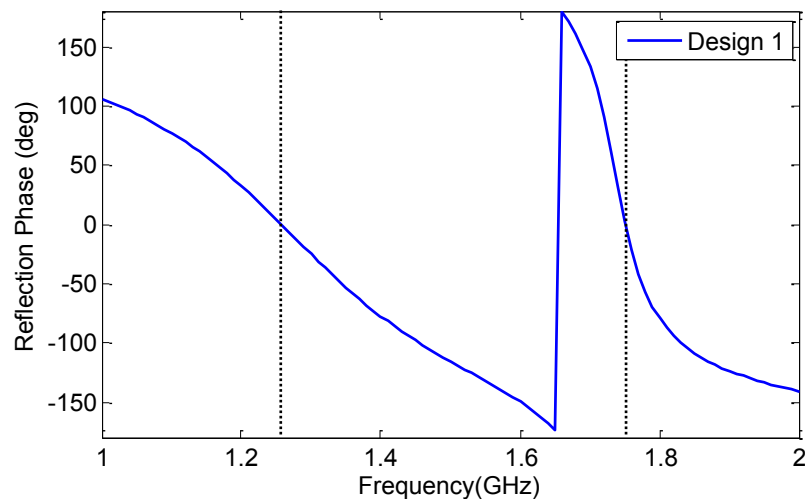
PEC Ground

$|E|$  Enhancement



1.25 GHz

1.75 GHz



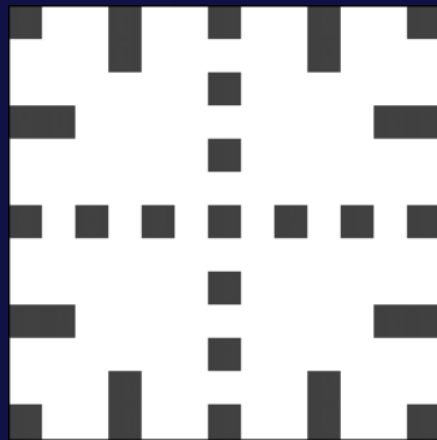


# Dual-Band AMC With Suppressed MFEF

Design Goal: AMC Condition at 1.25 GHz and 1.75 GHz

$$Cost = \sum_{freqs} \{\varphi_R - 0.0\}^2 + \{\max(|E|) - 0.0\}^2$$

$$freqs = \{1.25 \text{ GHz}, 1.75 \text{ GHz}\}$$



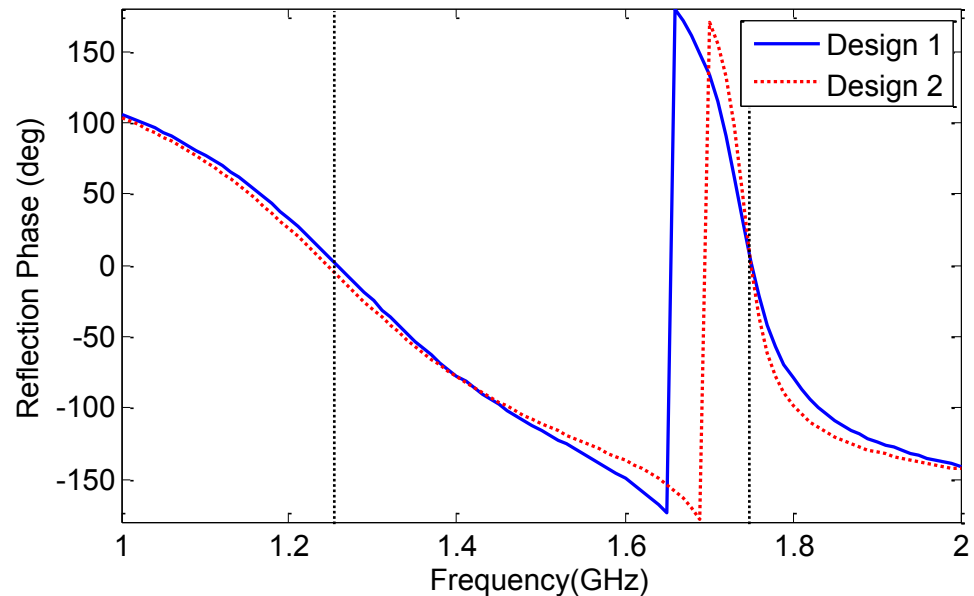
6.62 cm

FSS Screen

$\epsilon_r = 11.59$

PEC Ground

1.49 cm





# Dual-Band AMC MFEF Comparison

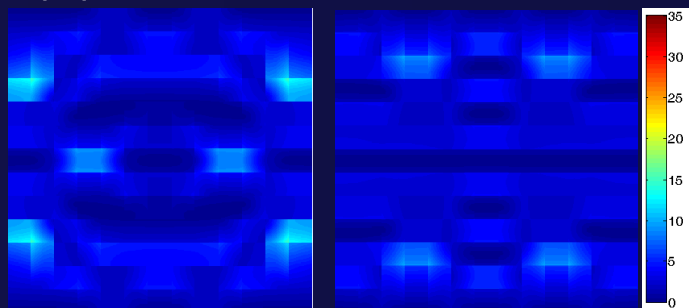
Peak MFEF @ 1.75 GHz Reduced from 31.0 to 15.8

Design 1

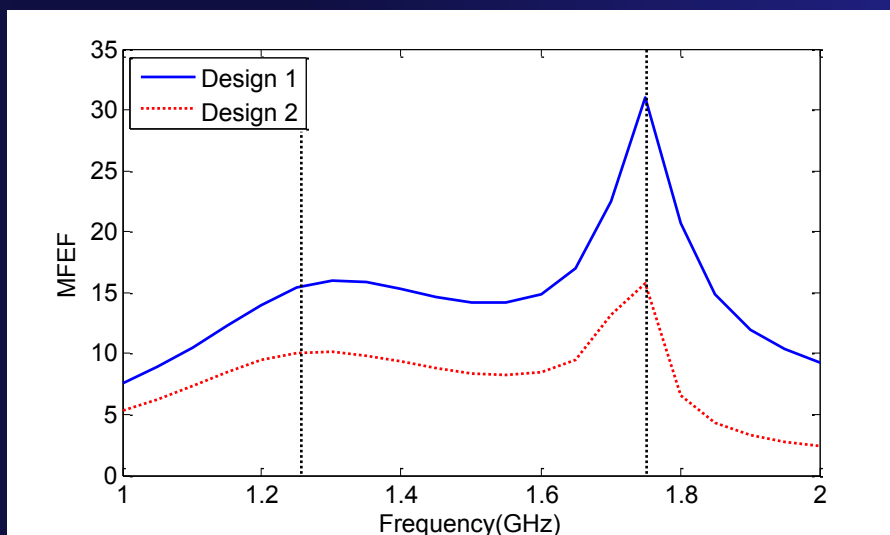
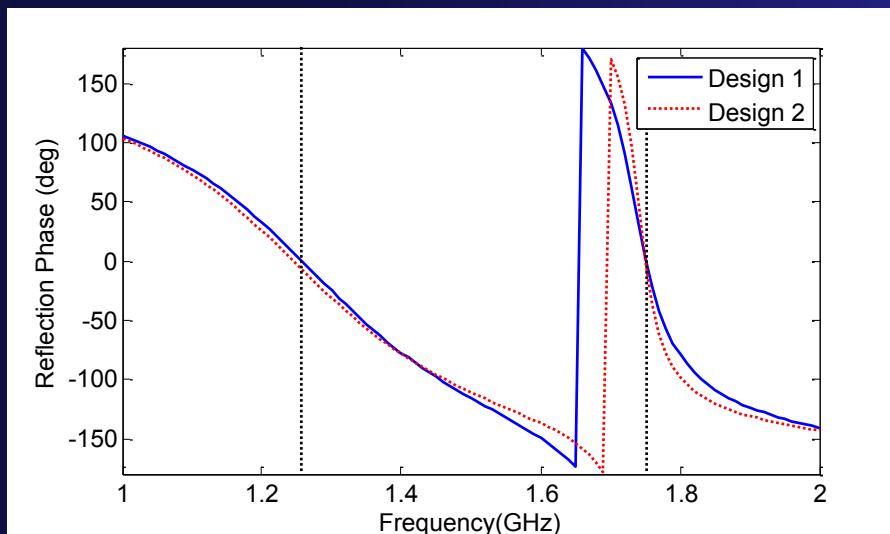
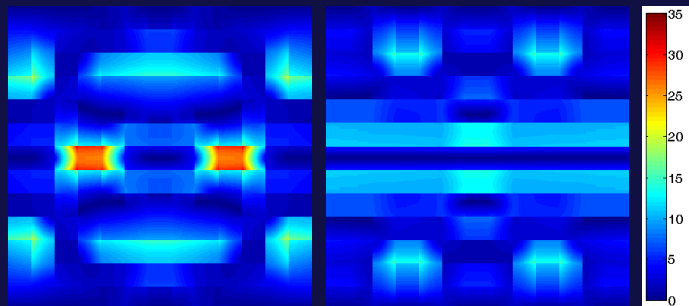
Design 2



$|E|$  Enhancement @ 1.25 GHz



$|E|$  Enhancement @ 1.75 GHz







# Summary

- Many metamaterial types rely on resonant behaviors that produce high fields within their structures.
- However, if a metamaterial can operate away from resonance (*e.g.*, low-index or zero-index metamaterials), it can be well-suited for HPM applications.
- Artificial Magnetic Conducting surfaces often exhibit high field enhancement at resonance with unoptimized MFEFs over 30.
- Genetic algorithm optimization was successfully employed to design single- and dual-band AMC surfaces with 50% reduced MFEF for HPM applications.